Low-Parametric-Sensitivity Realizations With Relaxed L_2 -Dynamic-Range-Scaling Constraints

Thibault Hilaire, Member, IEEE

Abstract—This brief presents a new dynamic-range scaling for the implementation of filters/controllers in state-space form. Relaxing the classical L_2 -scaling constraints by specific fixed-point considerations allows for a higher degree of freedom for the optimal L_2 -parametric sensitivity problem. However, overflows in the implementation are still prevented. The underlying constrained problem is converted into an unconstrained problem for which a solution can be provided. This leads to realizations that are still scaled but less sensitive.

Index Terms—Coefficient sensitivity, digital filter implementation, fixed-point implementation, scaling.

I. INTRODUCTION

T HE MAJORITY of control (or signal processing) systems is implemented in digital general-purpose processors, digital signal processors (DSPs), field-programmable gate arrays (FPGAs), etc. Since these devices cannot compute with infinite precision and approximate real-number parameters with a finite binary representation, the numerical implementation of controllers (filters) leads to deterioration in characteristics and performance. This has two separate origins, corresponding to the quantization of the embedded coefficients and the roundoff errors occurring during the computations. They can be formalized as parametric errors and numerical noises, respectively. The focus of this brief is on parametric errors, but one can refer to [1]–[4] for roundoff noises.

It is also well known that these finite-wordlength effects depend on the structure of the realization. This motivates us to investigate the minimization problem. It has widely been studied since Thiele published [5] and [6], and the definition of a tractable input–output sensitivity norm (the L_1/L_2 -sensitivity). This work has been extended with a more natural and reasonable measure, the L_2 -sensitivity [1], [7]. The dynamic-range-scaling constraints have been introduced in [8] and [9] to prevent overflow and underflow during the evaluation of the state vector and the state and criteria normalization. These constraints have to be considered in the L_2 -sensitivity minimization problem, for which Hinamoto *et al.* [10] propose an efficient quasi-Newton algorithm to solve it.

This brief investigates the L_2 -dynamic-range-scaling problem by considering the concrete fixed-point implementation of

Manuscript received October 15, 2008; revised February 24, 2009. Current version published July 17, 2009. This work was supported in part by the National Research Network "Signal and Information Processing in Science and Engineering" (NFN SISE) and in part by the Cairn Project and the Institut National de Recherche en Informatique et en Automatique, France. This paper was recommended by Associate Editor M. Chakraborty.

The author is with the Institute of Communications and Radio-Frequency Engineering, Vienna University of Technology, 1040 Vienna, Austria (e-mail: thibault.hilaire@nt.tuwien.ac.at).

Digital Object Identifier 10.1109/TCSII.2009.2022210

state-space realizations. It reveals that the classical L_2 -scaling is only a sufficient condition to prevent overflows, and thus, it can slightly be relaxed to extend the degrees of freedom for the optimization process. New *relaxed* L_2 -dynamic-range scalings are then presented with respect to the described computational scheme. Finally, the L_2 -sensitivity minimization problem with relaxed L_2 -scaling constraints is solved. A numerical example illustrates that the proposed constraints can offer reduced L_2 -sensitivity with overflow protection.

II. L_2 -Sensitivity Analysis

Let (A, b, c, d) be a stable, controllable, and observable linear discrete-time single-input-single-output state-space system, i.e.,

$$\begin{cases} \boldsymbol{x}(k+1) = \boldsymbol{A}\boldsymbol{x}(k) + \boldsymbol{b}\boldsymbol{u}(k) \\ \boldsymbol{y}(k) = \boldsymbol{c}\boldsymbol{x}(k) + d\boldsymbol{u}(k) \end{cases}$$
(1)

where $A \in \mathbb{R}^{n \times n}$, $b \in \mathbb{R}^{n \times 1}$, $c \in \mathbb{R}^{1 \times n}$, and $d \in \mathbb{R}$. u(k) is the scalar input, y(k) is the scalar output, and $x(k) \in \mathbb{R}^{n \times 1}$ is the state vector.

Its input–output relationship is given by the scalar transfer function $h : \mathbb{C} \to \mathbb{C}$ defined by

$$h: z \mapsto \boldsymbol{c}(z\boldsymbol{I}_n - \boldsymbol{A})^{-1}\boldsymbol{b} + d.$$
⁽²⁾

The quantization of the coefficients introduces some uncertainty to A, b, c, and d, leading to $A + \Delta A$, $b + \Delta b$, $c + \Delta c$, and $d + \Delta d$, respectively. It is of interest to consider the sensitivity of the transfer function with respect to the coefficients, based on the following definitions.

Definition 1 (Transfer Function Sensitivity): Consider $X \in \mathbb{R}^{m \times n}$ to be a matrix and $f : \mathbb{R}^{m \times n} \to \mathbb{C}$ to be a scalar complex function, differentiable with respect to all the entries of X. The sensitivity of f with respect to X is defined by the matrix $S_X \in \mathbb{R}^{m \times n}$

$$\frac{\partial f}{\partial \mathbf{X}} \stackrel{\Delta}{=} \mathbf{S}_{\mathbf{X}}, \quad \text{with} \quad (\mathbf{S}_{\mathbf{X}})_{i,j} \stackrel{\Delta}{=} \frac{\partial f}{\partial \mathbf{X}_{i,j}}. \quad (3)$$

Definition 2 (L_p -Norm): Let $H : \mathbb{C} \to \mathbb{C}^{k \times l}$ be a function of the scalar complex variable z. $||H||_p$ is the L_p -norm of H, defined by

$$\|\boldsymbol{H}\|_{p} \stackrel{\Delta}{=} \left(\frac{1}{2\pi} \int_{0}^{2\pi} \|\boldsymbol{H}(e^{j\omega})\|_{F}^{p} d\omega\right)^{\frac{1}{p}}$$
(4)

where $\|.\|_F$ is the Froebenius norm.

1549-7747/\$25.00 © 2009 IEEE

Gevers and Li [1] have proposed the L_2 -sensitivity measure to evaluate the coefficient roundoff errors. It is defined by

$$M_{L_2} \triangleq \left\| \frac{\partial h}{\partial \boldsymbol{A}} \right\|_2^2 + \left\| \frac{\partial h}{\partial \boldsymbol{b}} \right\|_2^2 + \left\| \frac{\partial h}{\partial \boldsymbol{c}} \right\|_2^2 + \left\| \frac{\partial h}{\partial \boldsymbol{d}} \right\|_2^2 \qquad (5)$$

and can be computed by $\frac{\partial h}{\partial \boldsymbol{A}}(z) = \boldsymbol{G}^{\top}(z)\boldsymbol{F}^{\top}(z), \ \frac{\partial h}{\partial \boldsymbol{b}}(z) = \boldsymbol{G}^{\top}(z), \ \frac{\partial h}{\partial \boldsymbol{b}}(z) = \boldsymbol{F}(z), \text{ and } \frac{\partial h}{\partial d}(z) = 1, \text{ with }$

$$\mathbf{F}(z) \stackrel{\Delta}{=} (z\mathbf{I}_n - \mathbf{A})^{-1} \mathbf{b} \quad \mathbf{G}(z) \stackrel{\Delta}{=} \mathbf{c}(zI_n - \mathbf{A})^{-1}.$$
 (6)

This measure is an extension of the more tractable but less natural L_1/L_2 -sensitivity measure proposed by Tavşanoğlu and Thiele [5] $[\|\frac{\partial h}{\partial A}\|_1^2$ instead of $\|\frac{\partial h}{\partial A}\|_2^2$ in (5)]. *Remark 1:* It is also possible to regroup all the coefficients

Remark 1: It is also possible to regroup all the coefficients in one unique matrix

$$\boldsymbol{Z} \stackrel{\Delta}{=} \begin{pmatrix} \boldsymbol{A} & \boldsymbol{b} \\ \boldsymbol{c} & \boldsymbol{d} \end{pmatrix}. \tag{7}$$

Then, with the L_2 -norm property, $M_{L_2} = \|\frac{\partial h}{\partial Z}\|_2^2$. From (6) and the associated state spaces, the sensitivity transfer function $\frac{\partial h}{\partial Z}$ can be described by the multiple-input–multiple-output (MIMO) state-space system $(\tilde{A}, \tilde{B}, \tilde{C}, \tilde{D})$ with

$$\tilde{A} \stackrel{\Delta}{=} \begin{pmatrix} A & bc \\ 0 & A \end{pmatrix} \quad \tilde{B} \stackrel{\Delta}{=} \begin{pmatrix} 0 & b \\ I_n & 0 \end{pmatrix} \\
\tilde{C} \stackrel{\Delta}{=} \begin{pmatrix} I_n & 0 \\ 0 & c \end{pmatrix} \quad \tilde{D} \stackrel{\Delta}{=} \begin{pmatrix} 0 & 0 \\ 0 & 1 \end{pmatrix}.$$
(8)

See [1] and [11] for more details.

The following proposition allows us to compute M_{L_2} .

Proposition 1: Let us consider H as the MIMO state-space system (K, L, M, N). Its L_2 -norm can be computed by

$$\|\boldsymbol{H}\|_{2}^{2} = \operatorname{tr}(\boldsymbol{N}\boldsymbol{N}^{\top} + \boldsymbol{M}\boldsymbol{W}_{c}\boldsymbol{M}^{\top})$$
(9)

$$= \operatorname{tr}(\boldsymbol{N}^{\top}\boldsymbol{N} + \boldsymbol{L}^{\top}\boldsymbol{W}_{o}\boldsymbol{L})$$
(10)

where W_c and W_o are the controllability and observability Gramians, respectively. They are the solutions of the Lyapunov equations

$$\boldsymbol{W}_{c} = \boldsymbol{K} \boldsymbol{W}_{c} \boldsymbol{K}^{\top} + \boldsymbol{L} \boldsymbol{L}^{\top} \quad \boldsymbol{W}_{o} = \boldsymbol{K}^{\top} \boldsymbol{W}_{o} \boldsymbol{K} + \boldsymbol{M}^{\top} \boldsymbol{M}.$$
(11)

Applying a coordinate transformation, defined by $\bar{\boldsymbol{x}}(k) \stackrel{\Delta}{=} \boldsymbol{T}^{-1}\boldsymbol{x}(k)$, to the state-space system $(\boldsymbol{A}, \boldsymbol{b}, \boldsymbol{c}, d)$ leads to a new equivalent realization $(\boldsymbol{T}^{-1}\boldsymbol{A}\boldsymbol{T}, \boldsymbol{T}^{-1}\boldsymbol{b}, \boldsymbol{c}\boldsymbol{T}, d)$.

Since these two realizations are equivalent in infinite precision but are not equivalent in finite precision (fixed-point arithmetic, floating-point arithmetic, etc.), the L_2 -sensitivity then depends on T and is denoted by $M_{L_2}(T)$.

In this case, it is natural to define the following problem:

Problem 1 (Optimal L_2 -Sensitivity Problem): Considering a state-space realization (A, b, c, d), the optimal L_2 -sensitivity problem consists of finding the coordinate transformation T_{opt} that minimizes M_{L_2}

$$\boldsymbol{T}_{\mathrm{opt}} = \mathop{\mathrm{arg\ min}}_{\boldsymbol{T} \mathrm{invertible}} M_{L_2}(\boldsymbol{T}).$$
 (12)

Reference [1] shows that the problem has one unique solution. Hence, for example, a gradient method can be used to solve it.

III. L_p -Dynamic-Range Scaling

The L_p -dynamic-range-scaling constraints have been introduced by Jackson in [8] and Hwang in [9]. They consist of scaling the state-variable vector such that overflows or underflows during its evaluation are prevented.

Definition 3 (L_p -Scaling): A state-space realization (A, b, c, d) is said to be L_p -scaled if the L_p -norms of the transfer functions from the input to each state are set to 1, i.e.,

$$\left\|\boldsymbol{e}_{i}^{\top}(\boldsymbol{z}\boldsymbol{I}_{n}-\boldsymbol{A})^{-1}\boldsymbol{b}\right\|_{p}=1\qquad\forall\,1\leqslant i\leqslant n\qquad(13)$$

where e_i is the column vector of appropriate dimension and with all elements being 0 except for the *i*th element, which is 1.

Let $\overset{\max}{u}$ denote the maximum value of the input u

$$\overset{\max\Delta}{u} = \max_{k \in \mathbb{N}} |u(k)|.$$
(14)

The L_1 -scaling guarantees that the dynamic of each state x_i is lower than u^{\max} , whereas the L_2 -scaling guarantees that the variance of each state is unitary for a unit-variance centered white noise input. L_2 -scaling does not completely prevent overflow as does L_1 , but it is less conservative and more realistic, so it is widely used [12].

With proposition 1 applied to the system $(\mathbf{A}, \mathbf{b}, \mathbf{e}_i^{\top}, \mathbf{0})$, the L_2 -scaling constraints (13) can be expressed as

$$(\boldsymbol{W}_c)_{i,i} = 1 \qquad \forall \, 1 \leqslant i \leqslant n \tag{15}$$

where W_c is the controllability Gramian of the state-space system (A, b, c, d).

Problem 2 (Sensitivity Problem With L_2 -Scaling Constraints): The optimal L_2 -sensitivity problem with L_2 -scaling constraints can be formulated as optimization problem 1, subject to the constraints in (15).

Moreover, it is possible to L_2 -scale a realization with the following proposition.

Proposition 2 (a posteriori L_2 -Scaling): Considering a state-space realization $(\mathbf{A}, \mathbf{b}, \mathbf{c}, d)$, it is also possible to a posteriori L_2 -scale it with a diagonal coordinate transformation T such that

$$\boldsymbol{T}_{i,i} = \sqrt{(\boldsymbol{W}_c)_{i,i}} \qquad \forall \, 1 \leqslant i \leqslant n.$$
 (16)

Then, there exist infinite transformation matrices T (not necessarily diagonal) that produces L_2 -scaled realizations: let us consider the invertible matrix $U \in \mathbb{R}^{n \times n}$; then, the transformation matrix T = UV also produces L_2 -scaling with V diagonal such that

$$\boldsymbol{V}_{i,i} = \sqrt{(\boldsymbol{U}^{-1} \boldsymbol{W}_c \boldsymbol{U}^{-\top})_{i,i}} \qquad \forall 1 \leq i \leq n.$$
(17)

Proof: A transformation matrix T that transforms (A, b, c, d) into $(T^{-1}AT, T^{-1}b, cT, d)$ changes the controllability Gramian W_c into $T^{-1}W_cT^{-\top}$.

Since T is diagonal, the constraints $(W_c)_{i,i} = 1$ imply that $T_{i,i} = \sqrt{(W_c)_{i,i}}$.

Moreover, it is also possible to successively apply two transformation matrices U and V on (A, b, c, d). If V is composed according to (17), then the transformation T = UV performs the L_2 -scaling.

This proposition can be used to transform constrained problem 2 into an unconstrained problem. Then, an



Fig. 1. Fixed-point representation.

optimization algorithm like quasi-Newton can be used to solve it. Other analytical algorithms for this problem can be found in [3] and [10].

IV. FIXED-POINT IMPLEMENTATION

A. Fixed-Point Representation

In this brief, the notation (β, γ) is used for the fixed-point representation of a variable or coefficient (2's complement scheme), according to Fig. 1. β is the total wordlength of the representation in bits, whereas γ is the wordlength of the fractional part (it determines the position of the binary point). They are fixed for each variable (input, states, and output) and each coefficient and implicit (unlike the floating-point representation). β and γ will be suffixed by the variable/coefficient they refer to.

To represent a value x without overflow, a fixed-point representation (β_x, γ_x) may satisfy

$$\beta_x - \gamma_x - 1 \ge \lfloor \log_2 |x| \rfloor + 1 \tag{18}$$

where the $\lfloor a \rfloor$ operation rounds *a* to the nearest integer less than or equal to *a* (for positive numbers $\lfloor a \rfloor$ is the integer part).

An important fixed-point issue is to find a valid fixed-point representation such that (18) is satisfied for all values that x can assume during the execution of the algorithm.

B. State Overflow

Definition 4 (State Overflow): The overflow of the state variables $(x_i)_{1 \le i \le n}$ can be strictly avoided iff $(1 \le i \le n)$

$$\forall k, \qquad -2^{\beta_{x_i} - \gamma_{x_i} - 1} \leqslant \boldsymbol{x}_i(k) < 2^{\beta_{x_i} - \gamma_{x_i} - 1}. \tag{19}$$

The overflows are avoided if the binary-point position of each state is carefully chosen such that

$$\gamma_{\boldsymbol{x}_i} = \beta_{\boldsymbol{x}_i} - 2 - \left\lfloor \log_2 \frac{\max}{\boldsymbol{x}_i} \right\rfloor \tag{20}$$

where $\overset{\max}{m{x}_i}$ is the maximum magnitude for the ith state

$$\mathbf{x}_{i}^{\max\Delta} = \max_{k \in \mathbb{N}} |\mathbf{x}_{i}(k)|.$$
 (21)

However, only upper bounds can be computed. The first upper bound x_i^{up} can be obtained by an L_1 -norm

$$\overset{\text{up}}{\boldsymbol{x}_{i}} = \left\| \boldsymbol{e}_{i}^{\top} (\boldsymbol{z} \boldsymbol{I}_{n} - \boldsymbol{A})^{-1} \boldsymbol{b} \right\|_{1} \overset{\text{max}}{\boldsymbol{u}}$$
(22)

and the second one can be estimated by an L_2 -norm [12]

$$\mathbf{x}_{i}^{\mathrm{up}} \simeq \delta \left\| \boldsymbol{e}_{i}^{\top} (\boldsymbol{z} \boldsymbol{I}_{n} - \boldsymbol{A})^{-1} \boldsymbol{b} \right\|_{2} \overset{\mathrm{max}}{\boldsymbol{u}} .$$
 (23)

Here, the parameter δ can be interpreted as a representation of the number of standard deviations of x_i if the input is unit-

variance white centered noise $(\delta \ge 1)$. Since the L_2 -norm in (23) does not give a strict bound [contrary to (22)], δ can be seen as a *safety* parameter [12].

Finally, these upper bounds are used to define the binarypoint positions

$$\gamma_{\boldsymbol{x}_i} = \beta_{\boldsymbol{x}_i} - 2 - \left\lfloor \log_2 \boldsymbol{x}_i^{\text{up}} \right\rfloor.$$
 (24)

In general, the L_1 and L_2 estimations of \mathbf{x}_i^{up} approximately leads to the same binary-point position, with 1- or 2-bit deviation. However, since the L_2 -norm is more tractable (with proposition 1) and the L_1 -norm is too conservative $(\overset{\text{max}}{x_i} \ll \overset{\text{up}}{x_i})$, in practice, (23) is used, with $\delta = 1$. After implementation, a simulation-based estimation like in [13] or [14] can also be used to verify *in situ* the peak values and the binary-point positions, according to the inputs.

C. Computational Scheme

To implement a realization without overflows, two equivalent choices are possible.

- Set the binary-point position for each state according to (24) to make sure that the fixed-point representation of the states avoids state overflows.
- Define a binary-point position for each state and apply a scaling to them to adapt the peak values of each state to the chosen binary-point position.

Here, we here focus on the second choice, referring to dynamicrange-scaling constraints.

Let us consider in detail the fixed-point implementation of the system given in (1). It leads to (n + 1) scalar products to be evaluated of the form

$$S = \sum_{i=1}^{N} \boldsymbol{p}_i \boldsymbol{q}_i \tag{25}$$

where the (p_i) are the given coefficients, and (q_i) are bounded variables.

To avoid bit-shift operations between each addition in the evaluation of (25), the binary-point positions of each partial product of the sum should be equal.

Then, two computational schemes are possible: the *roundoff*after-multiplication scheme, where shifts are added after each product to align the operands of the sum ($p_i q_i$ is implemented as ($p'_i * q'_i$) >> d_i), and the *roundoff-before-multiplication* scheme, where the required shifts are reported into the coefficients ($p_i q_i$ is implemented as (p'_i >> d_i) * q'_i).

The main idea of the scaling is to scale each variable (q_i) such that the shifts $(d_i = 0, \forall i)$ are prevented. In fixed-point representation, the scaling only implies that all the (q_i) have a common format, and so do all the (p_i) . See [2] and [15] for more details on implementation schemes.

Applied to the state-space realization (1), this yields that all the states must have the same binary-point position as the input and the coefficients A, b, c, and d.

In addition, since they have the same fractional part, their quantization's errors ΔA , Δb , Δc , and Δd have the same

magnitude $2^{-\gamma z-1}$, and the L_2 -sensitivity measure represents a meaningful bound on the transfer function error Δh

$$\begin{split} \|\Delta h\|_{2}^{2} &\leqslant \left\|\frac{\partial h}{\partial \boldsymbol{A}} \times \boldsymbol{\Delta} \boldsymbol{A}\right\|_{2}^{2} + \left\|\frac{\partial h}{\partial \boldsymbol{b}} \times \boldsymbol{\Delta} \boldsymbol{b}\right\|_{2}^{2} \\ &+ \left\|\frac{\partial h}{\partial \boldsymbol{c}} \times \boldsymbol{\Delta} \boldsymbol{c}\right\|_{2}^{2} + \left\|\frac{\partial h}{\partial \boldsymbol{d}} \times \boldsymbol{\Delta} \boldsymbol{d}\right\|_{2}^{2} \qquad (26) \\ &\leqslant 2^{-2(\gamma_{Z}+1)} M_{L_{2}}. \end{split}$$

D. New L₂-Scaling Constraints

Taking this into consideration, the overflows will be avoided by setting the same binary-point position for the states and the input and by applying an appropriate scaling on the states such that the constraints (20) are satisfied.

Compared to strict L_2 -scaling where the states must satisfy $\mathbf{x}_i^{\max} = \mathbf{u}^{\max}$, here, the constraints are relaxed (but still restrictive enough to guarantee the protection against overflow) and replaced by $\gamma_{\mathbf{x}_i} = \gamma_u$.

Proposition 3 (Relaxed L_2 -Scaling Constraints): Since the input and the states may have the same binary-point position, the L_2 -scaling constraints (15) are now transformed into

$$\frac{2^{2\boldsymbol{\alpha}_i}}{\delta^2} \leqslant (\boldsymbol{W}_c)_{i,i} < 4\frac{2^{2\boldsymbol{\alpha}_i}}{\delta^2} \qquad \forall 1 \leqslant i \leqslant n$$
(28)

where

$$\boldsymbol{\alpha}_{i} \stackrel{\Delta}{=} \beta_{\boldsymbol{x}_{i}} - \beta_{u} - \mathcal{F}_{2} \begin{pmatrix} \max \\ u \end{pmatrix}$$
(29)

and $\mathcal{F}_2(x)$ is defined as the fractional value of $\log_2(x)$

$$\mathcal{F}_2(x) \stackrel{\Delta}{=} \log_2(x) - \lfloor \log_2(x) \rfloor \,. \tag{30}$$

For microcontroller or DSP implementations (contrary to FPGA or some application-specified integrated circuit implementations), the wordlength of all variables is equal, i.e., $\beta_u = \beta_{\boldsymbol{x}_i}$ $(1 \leq i \leq n)$. Furthermore, $\overset{\text{max}}{u}$ could be set to a power of two. Then, if δ is set to unity (as for classical L_2 -scaling constraints), the relaxed L_2 -scaling constraints (28) become

$$1 \leqslant (\boldsymbol{W}_c)_{i,i} < 4 \qquad \forall 1 \leqslant i \leqslant n.$$
(31)

Proof: The binary-point position of the input is set to $\gamma_u = \beta_u - 2 - \lfloor \log_2 \frac{\max}{u} \rfloor$. Hence, with (24), the constraints $\gamma_u = \gamma_{\boldsymbol{x}_i}$ lead to

$$\beta_{u} - \lfloor \log_{2} \overset{\max}{u} \rfloor = \beta_{\boldsymbol{x}_{i}} - \left\lfloor \log_{2} \left(\delta \left\| \boldsymbol{e}_{i}^{\mathsf{T}} (\boldsymbol{z} \boldsymbol{I}_{n} - \boldsymbol{A})^{-1} \boldsymbol{b} \right\|_{2} \overset{\max}{u} \right) \right\rfloor$$

and

$$\left\lfloor \log_2 \left(\delta \sqrt{(\boldsymbol{W}_c)_{i,i}} \right) + \mathcal{F}_2(\overset{\max}{\boldsymbol{u}}) \right\rfloor = \beta_{\boldsymbol{x}_i} - \beta_u \qquad (32)$$

and finally

$$2^{\boldsymbol{\alpha}_i} \leqslant \delta \sqrt{(\boldsymbol{W}_c)_{i,i}} < 2^{\boldsymbol{\alpha}_i+1}.$$
(33)

It is important to remark that these new constraints allow more freedom for the scaling and introduce a new degree of freedom for the search for optimal realizations. Moreover, even though not considered in this brief, it could give more freedom for the minimization of the roundoff noise power.

V. Optimal L_2 -Sensitivity Realization With Relaxed L_2 -Norm Dynamic-Range-Scaling Constraints

Then, these relaxed constraints can be applied to a new sensitivity problem:

Problem 3 (Relaxed Sensitivity Problem): The optimal L_2 sensitivity problem with relaxed L_2 -norm dynamic-rangescaling constraints can be expressed in the form of constrained problem 2 subject to the constraints in (28).

This constrained problem can be solved by two different means.

First, in addition to the n^2 free parameters of the transformation matrix U applied to the system, n extra parameters $(\gamma_i)_{1 \leq i \leq n}$ can be considered. These (γ_i) represent the desired L_2 -scaling and will be constrained by

$$\frac{2^{2\alpha_i}}{\delta^2} \leqslant \gamma_i < 4\frac{2^{2\alpha_i}}{\delta^2} \qquad \forall 1 \leqslant i \leqslant n.$$
(34)

Then, a diagonal transformation matrix V_{γ} is applied, with

$$(V_{\gamma})_{i,i} = \sqrt{\frac{(W_c)_{i,i}}{\gamma_i}}.$$
 (35)

In this case, a constrained optimization algorithm (like a quasi-Newton one, implemented in fmincon with Matlab) can then be used to solve the following problem:

$$(\boldsymbol{U}_{\text{opt}}, \gamma_{\text{opt}}) = \arg \min_{\substack{\boldsymbol{U} \text{ invertible} \\ \boldsymbol{\gamma} \text{ satisfying (34)}}} M_{L_2}(\boldsymbol{U}\boldsymbol{V}_{\boldsymbol{\gamma}}).$$
(36)

The optimal realization satisfying the relaxed L_2 constraints is then obtained by applying the transformation matrix $T_{opt} = U_{opt}V_{\gamma_{opt}}$.

The other approach is to scale the system after each transformation to ensure that the constraints are met:

Proposition 4 (a posteriori Relaxed Scaling): Considering a state-space realization, it is possible to a posteriori scale it with a diagonal transformation matrix T given by

$$\boldsymbol{T}_{i,i} = \delta \sqrt{(\boldsymbol{W}_c)_{i,i}} 2^{-\mathcal{F}_2\left(\delta \sqrt{(\boldsymbol{W}_c)_{i,i}}\right) - \boldsymbol{\alpha}_i}$$
(37)

such that the constraints (28) are satisfied. Moreover, it is possible to build all the transformation matrices that meet the constraints (28): let us consider an invertible matrix $U \in \mathbb{R}^{n \times n}$; then, the transformation matrix T = UV with V diagonal such that

$$\boldsymbol{V}_{i,i} = \delta \sqrt{(\boldsymbol{U}^{-1} \boldsymbol{W}_c \boldsymbol{U}^{-\top})_{i,i}} 2^{-\mathcal{F}_2 \left(\delta \sqrt{(\boldsymbol{U}^{-1} \boldsymbol{W}_c \boldsymbol{U}^{-\top})_{i,i}}\right) - \boldsymbol{\alpha}_i}$$
(38)

produces the relaxed L_2 -scaling.

Proof: \mathcal{F}_2 acts as a modulo operator. For $x \in \mathbb{R}$, $\bar{x} \stackrel{\Delta}{=} 2^{\mathcal{F}_2(x)+a}$ is such that $2^a \leq \bar{x} < 2^{a+1}$.

Since the constraints (28) are equal to

$$2^{\boldsymbol{\alpha}_i} \leqslant \delta \sqrt{(\boldsymbol{W}_c)_{i,i}} < 2^{\boldsymbol{\alpha}_i + 1} \tag{39}$$

and T transforms $(W_c)_{i,i}$ into $T_{i,i}^{-2}(W_c)_{i,i}$, then $T_{i,i}$ has to be of the form

$$\delta \boldsymbol{T}_{i,i}^{-1} \sqrt{(\boldsymbol{W}_c)_{i,i}} = 2^{\mathcal{F}_2\left(\delta\sqrt{(\boldsymbol{W}_c)_{i,i}}\right) + \boldsymbol{\alpha}_i}.$$
 (40)

Authorized licensed use limited to: Universitatsbibliothek der TU Wien. Downloaded on August 3, 2009 at 08:06 from IEEE Xplore. Restrictions apply.

TABLE I M_{L_2} -Sensitivities for the Realizations $\mathcal{R}_1, \mathcal{R}_2$ and \mathcal{R}_3

| realization | M_{L_2} sensitivity |
|-----------------|-----------------------|
| \mathcal{R}_1 | 6355.5 |
| \mathcal{R}_2 | 530.0964 |
| \mathcal{R}_3 | 528.2532 |

Thus, the optimization problem is given by

$$U_{\text{opt}} = \arg \min_{\substack{\boldsymbol{U} \text{ invertible} \\ \boldsymbol{V} \text{ defined by (38)}}} M_{L_2}(\boldsymbol{U}\boldsymbol{V}). \tag{41}$$

These two ways of solving problem 3 are implemented in the FWR toolbox¹ for Matlab, with fminsearch, fmincon, and fminunc functions, and they both give the same results with similar numbers of iterations.

Of course, the use of matrices V_{γ} and V, which are merely used to eliminate the constraints and solve an unconstrained minimization problem, increases the degree of nonlinearity for the objective function to minimize. However, this seems not to be a problem, since in our tests, the optimal realizations found seem to be global optima.

VI. EXAMPLE

Let us consider the following state-space digital controller, given in modal form²:

$$\boldsymbol{A} = \begin{pmatrix} 0.3820 & 0 & 0\\ 0 & 0.7964 & 0.5598\\ 0 & -0.5598 & 0.7964 \end{pmatrix} \boldsymbol{b} = \begin{pmatrix} 0.5391\\ -0.8417\\ 0.6232 \end{pmatrix}$$

$$\boldsymbol{c} = (0.1664 & 0.1639 & 0.2047) \quad \boldsymbol{d} = 0.0159 \quad (42)$$

and its multiple equivalent (in infinite-precision) realizations:

- 1) \mathcal{R}_1 is the original realization given by (42).
- 2) \mathcal{R}_2 is the optimal L_2 -scaled realization (solution of problem 2). It is obtained with proposition 2 and a quasi-Newton algorithm.
- 3) \mathcal{R}_3 is the optimal relaxed L_2 -scaled realization (problem 3), with $\overset{\text{max}}{u}$ being a power of two and $\delta = 1$. It is obtained with proposition 4.

Table I gives the M_{L_2} sensitivities of these different realizations.

In this example, the relaxed L_2 -scaled realization \mathcal{R}_3 achieves lower sensitivity than the strict L_2 -scaled optimal realization \mathcal{R}_2 while protecting implementation from overflows. However, it is not always the case: if we consider the example in [10], the optimal relaxed L_2 -scaled realization satisfies $(\mathbf{W}_c)_{i,i} = 1$ and is then also a strict L_2 -scaled realization. This depends on the diagonal terms of the controllability Gramians of the (nonscaled) optimal realization.

It is also interesting to notice that a good estimation of $\overset{\text{max}}{u}$ (if it is not a power of two) can allow achieving lower sensitivity by moving the constraints (it could also be the case for the example in [10]).

VII. CONCLUSION

This brief has presented the L_2 -sensitivity minimization problem and the associated L_2 -scaling constraints. These constraints that prevent overflows have been considered with concrete fixed-point implementation schemes. Novel L_2 -dynamic-range constraints have been exhibited.

Even if the goal of this brief is not a detailed optimization algorithm like in [10], two different means to solve the constrained optimization problem have been exhibited and applied on a numerical example.

These relaxed constraints could also be very important for some other realizations, like the δ -operator state space or the ρ -direct Form II transposed [16]. For these realizations where a parameter Δ should be used to achieve the L_2 -scaling, a relaxed- L_2 -scaling permits fixing this parameter as a power of two to decrease the amount of computations.

To apply this work to other classical structures, it will be soon extended to the specialized implicit framework [17] that allows to encompass existing structures in an implicit state-space form.

REFERENCES

- M. Gevers and G. Li, Parametrizations in Control, Estimation and Filtering Problems. New York: Springer-Verlag, 1993.
- [2] T. Hilaire, D. Ménard, and O. Sentieys, "Bit accurate roundoff noise analysis of fixed-point linear controllers," in *Proc. IEEE Int. Symp. CACSD*, Sep. 2008, pp. 607–612.
- [3] S. Hwang, "Minimum uncorrelated unit noise in state-space digital filtering," *IEEE Trans. Acoust., Speech, Signal Process.*, vol. ASSP-25, no. 4, pp. 273–281, Aug. 1977.
- [4] C. Mullis and R. Roberts, "Synthesis of minimum roundoff noise fixed point digital filters," *IEEE Trans. Circuits Syst.*, vol. CAS-23, no. 9, pp. 551–562, Sep. 1976.
- [5] V. Tavşanoğlu and L. Thiele, "Optimal design of state-space digital filters by simultaneous minimization of sensibility and roundoff noise," *IEEE Trans. Acoust., Speech, Signal Process.*, vol. CAS-31, no. 10, pp. 884– 888, Oct. 1984.
- [6] L. Thiele, "Design of sensitivity and round-off noise optimal state-space discrete systems," *Int. J. Circuit Theory Appl.*, vol. 12, no. 1, pp. 39–46, Jan. 1984.
- [7] T. Hinamoto and Y. Sugie, "L2-sensitivity analysis and minimization of 2-d separable-denominator state-space digital filters," *IEEE Trans. Signal Process.*, vol. 50, no. 12, pp. 3107–3114, Dec. 2002.
- [8] L. Jackson, "Roundoff-noise analysis for fixed-point digital filters realized in cascade or parallel form," *IEEE Trans. Audio Electroacoust.*, vol. AU-18, no. 2, pp. 107–122, Jun. 1970.
- [9] S. Hwang, "Dynamic range constraint in state-space digital filtering," *IEEE Trans. Acoust., Speech, Signal Process.*, vol. ASSP-23, no. 6, pp. 591–593, Dec. 1975.
- [10] T. Hinamoto, H. Ohnishi, and W.-S. Lu, "Minimization of L₂ sensitivity of one- and two dimensional state-space digital filters subject to L₂dynamic-range-scaling constraints," *IEEE Trans. Circuits Syst. II, Exp. Briefs*, vol. 52, no. 10, pp. 641–645, Oct. 2005.
- [11] T. Hilaire and P. Chevrel, "On the compact formulation of the derivation of a transfer matrix with respect to another matrix," INRIA, Paris, France, Tech. Rep. RR-6760, 2008.
- [12] K. Parhi, VLSI Digital Signal Processing Systems: Design and Implementation of Digital Controllers. Hoboken, NJ: Wiley, 1999.
- [13] P. Belanovic and M. Rupp, "Automated floating-point to fixed-point conversion with the Fixify environment," in *Proc. 16th IEEE Int. Workshop Rapid Syst. Prototyping*, Jun. 2005, pp. 172–178.
- [14] S. Kim, K. Kum, and W. Sung, "Fixed-point optimization utility for C and C++ based digital signal processing programs," *IEEE Trans. Circuits Syst.*, vol. 45, no. 11, pp. 1455–1464, Nov. 1998.
- [15] G. Constantinides, P. Cheung, and W. Luk, Synthesis and Optimization of DSP Algorithms. Norwell, MA: Kluwer, 2004.
- [16] G. Li and Z. Zhao, "On the generalized DFIIt structure and its state-space realization in digital filter implementation," *IEEE Trans. Circuits Syst.*, vol. 51, no. 4, pp. 769–778, Apr. 2004.
- [17] T. Hilaire, P. Chevrel, and J. Whidborne, "A unifying framework for finite wordlength realizations," *IEEE Trans. Circuits Syst.*, vol. 8, no. 54, pp. 1765–1774, Aug. 2007.

¹Sources are available at http://fwrtoolbox.gforge.inria.fr/

²Due to lack of space, only four digits are given, but more may be required to completely define the system.